Limits of Transmit Beamforming for Massive MIMO Radar

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Beamforming in Massive MIMO



Figure: Co-located MIMO radar.

Beamforming in Massive MIMO



Figure: Typical beampattern in a radar system.

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As the number of antennas, N, grows large

- What beampatterns can be realized if the covariance matrix of the transmit signals may be chosen at will?
- Output the seam of the seam
- How our ability to form a peak in a beampattern is governed by the number of antennas?

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Formulation

An array of N Tx antennas transmitting $x_n(l) \in \mathbb{C}$ The base band signal:

$$\sum_{n=1}^{N} e^{-j2\pi f_0 \tau_n(\theta)} x_n(l) \triangleq \mathbf{a}^H(\theta) \mathbf{x}(l), \quad l \in \{1, 2, \cdots, L\}$$

The power of the probing signal (transmit beampattern) at θ :

$$p(heta) = \mathbf{a}^{H}(heta)\mathbf{Ra}(heta)$$

where $\mathbf{R} = \mathbb{E}\{\mathbf{x}(l)\mathbf{x}^{H}(l)\}$ is the signal covariance matrix. Assume $\tau_{n}(\theta) = n\theta$ for ULA and $\xi = 2\pi f_{0} = 1$,

$$p(\theta) = \sum_{k=1}^{N} \sum_{l=1}^{N} R_{k,l} e^{j(k-l)\theta}$$

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Note that

A covariance matrix **R** of size $N \times N$ can always be realized with N independent streams of signals, transmitted by N antennas.

 Q: What functions $p(\theta)$ can be realized using *N* antennas, if the covariance matrix $\mathbf{R} \in \mathbb{C}^{N \times N}$ may be chosen at will?

Definition

The Zero-Order Resolution, $I_0(N)$ is defined as the number of points in space for which we can exactly determine the power, i.e., we can design the covariance matrix of the signal transmitted by N antennas in order to achieve the allocated power.

When $\|\mathbf{R}\|_F$ is bounded. Note that

$$p(\theta) = \mathbf{a}^{H}(\theta)\mathbf{R}\mathbf{a}(\theta) = \mathrm{tr}(\mathbf{R}\mathbf{a}(\theta)\mathbf{a}^{H}(\theta)) = \mathrm{tr}(\mathbf{R}\mathbf{\bar{A}}(\theta))$$

Let
$$\mathbf{E} = ar{\mathbf{A}}(heta_2) - ar{\mathbf{A}}(heta_1)$$
 for $heta_1$ and $heta_2$. Then

- $|p(\theta_2) p(\theta_1)| = |tr(\mathsf{RE})|$ will be small for a small $|\theta_2 \theta_1|$
- **②** Given a smoothness of $p(\theta)$, an N^2 point realization of the beampattern is achievable.

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The psd constraint can be equivalently expressed as $\mathbf{R} = \mathbf{X}^H \mathbf{X}$ for any $\mathbf{X} \in \mathbb{C}^{N \times N}$. Then $\|\mathbf{X}\mathbf{a}(\theta)\|_2 = \sqrt{p(\theta)}$. Suppose the beam pattern $p(\theta)$ is to be realized at N locations $\{\theta_k\}_{k=1}^N$ and thus $\mathbf{X}\mathbf{a}(\theta_k) = \sqrt{p(\theta_k)}\mathbf{u}_k$. Let

$$\begin{aligned} \mathbf{A} &= [\mathbf{a}(\theta_1) \ \mathbf{a}(\theta_2) \ \cdots \ \mathbf{a}(\theta_N)], \\ \mathbf{U} &= [\mathbf{u}_1 \ \mathbf{u}_2 \ \cdots \ \mathbf{u}_N], \\ \mathbf{D} &= \text{Diag}\left(\left[\sqrt{p(\theta_1)} \ \sqrt{p(\theta_2)} \ \cdots \ \sqrt{p(\theta_N)}\right]\right). \end{aligned}$$

Note that A is a non-singular Vandermonde matrix. Then

 $\begin{array}{l} \mathsf{X}\mathsf{A} &= \mathsf{U}\mathsf{D} \\ \Rightarrow & \mathsf{X} &= \mathsf{U}\mathsf{D}\mathsf{A}^{-1} \\ \Rightarrow & \mathsf{R} &= \mathsf{A}^{-H}\mathsf{D}\mathsf{U}^{H}\mathsf{U}\mathsf{D}\mathsf{A}^{-1} \end{array}$

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Q: How rapidly we can change the beampattern, $p(\theta)$ for closely located angles θ using N antennas?

Note that,

$$\frac{\partial p(\theta)}{\partial \theta} = \sum_{k,l} j(k-l) R_{k,l} e^{j(k-l)\theta},$$

which implies

$$\left|\frac{\partial p(\theta)}{\partial \theta}\right| \leq 2 \sum_{k>l} (k-l) |R_{k,l}| \triangleq l_1(\mathbf{R})$$

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$$\frac{|I_1(\mathbf{R})|^2}{\|\mathbf{R}\|_F^2} \leq \frac{\alpha}{6} N^2 (N^2 - 1)$$

where $\alpha = \frac{\|\mathbf{R}\|_F^2 - R_{\text{diag}}}{\|\mathbf{R}\|_F^2}$.

Theorem

Assuming that the transmission power is fixed with respect to the number of antennas, N, the rate of innovation $I_1(N)$ behaves as $\mathcal{O}(N^2)$ with respect to N.

Q: How our ability to form a peak in a beampattern is governed by the number of antennas?

To form a peak one must be able to make the second derivative of $p(\theta)$ "large":

$$\frac{\partial^2 p(\theta)}{\partial \theta^2} = \sum_{k,l} -(k-l)^2 R_{k,l} e^{j(k-l)\theta}$$

which implies that

$$\left|\frac{\partial^2 p(\theta)}{\partial \theta}\right| \leq 2 \sum_{k>l} (k-l)^2 |R_{k,l}| \triangleq l_2(\mathbf{R})$$

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It follows from above

$$\frac{|I_2(\mathbf{R})|^2}{\|\mathbf{R}\|_F^2} = \mathcal{O}(N^6)$$

Theorem

Assuming that the transmission power is fixed with respect to the number of antennas, N, forming of a peak in the beampattern, $I_2(N)$ behaves as $\mathcal{O}(N^3)$.

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Mathematically, the beampattern matching is accomplished by solving the following problem,

$$\min_{\zeta, \mathbf{R}} \frac{1}{K} \sum_{k=1}^{K} \omega_k [\mathbf{a}^H(\theta_k) \mathbf{R} \mathbf{a}(\theta_k) - \zeta d(\theta_k)]^2$$
s.t. $R_{n,n} = c/N, \ n = 1, \cdots, N,$
 $\mathbf{R} \succeq \mathbf{0}$

where

$$d(heta) = \left\{ egin{array}{ccc} 1, & heta \in [ilde{ heta}_k - rac{ riangle}{2}, ilde{ heta}_k + rac{ riangle}{2}], & k = 1, 2, 3, \cdots \ 0, & ext{otherwise} \end{array}
ight.$$

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Numerical Studies (contd.)



Figure: Realization of a smooth sinusoidal beampattern with zero-order resolution K = 181 using $N \in \{3, 6, 9, 12, 15, 18\}$ antennas.

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Numerical Studies (contd.)



Figure: Realization of a rectangular beampattern with resolution K = 181 using $N \in \{5, 15, 25, 35, 45\}$ antennas.

Numerical Studies (contd.)



Figure: Realization of a impulse-like beampattern with resolution K = 181 using $N \in \{10, 25, 50, 100, 125\}$ antennas.

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• The fundamental limitations of the resolution of beampatterns produced by MIMO radars in relation to their number of antennas.

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- Multiple analytical results to show how the changes in a beampattern are impacted by an increased number of antennas in a massive MIMO scenario.
- Future research: The characterization and efficient construction of such beampatterns.

Thank you and Questions?

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