

Deep Radar Waveform Design for Efficient Automotive Radar Sensing

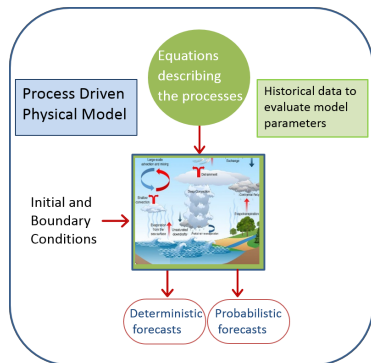
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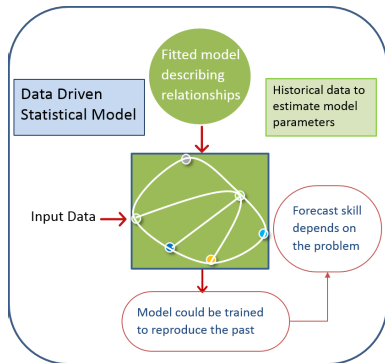
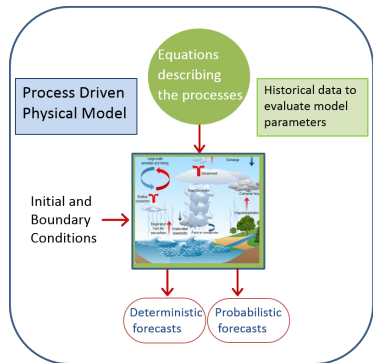
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- 1 Motivation
- 2 Problem Formulation
- 3 Numerical Examples
- 4 Conclusion

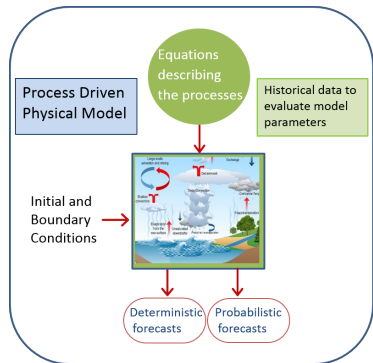
Model-Based and Data-Driven Approaches



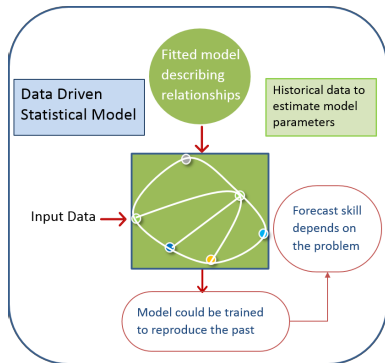
Model-Based and Data-Driven Approaches



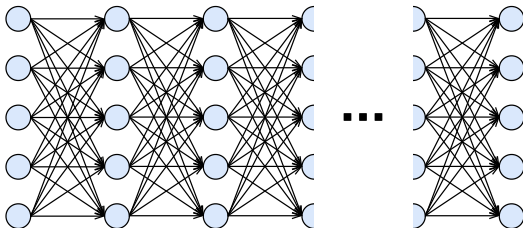
Model-Based and Data-Driven Approaches



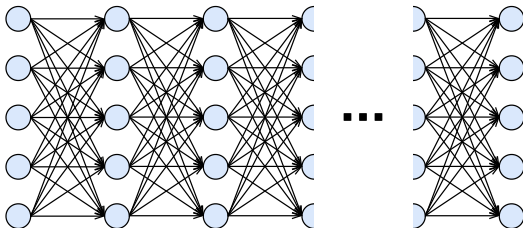
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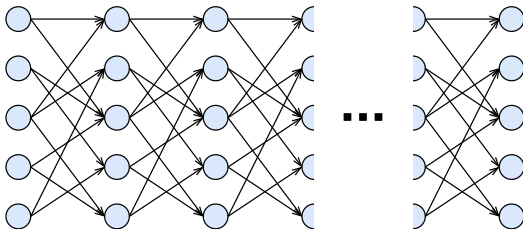
Deep Neural
Network (DNN)



Deep Neural Network (DNN)



Deep Unfolding Network (DUN)



Sequence:

$$\mathbf{s} = [s_1 \ s_2 \ \cdots \ s_N]^T \in \mathbb{C}^N$$

Output:

$$\mathbf{y} = \mathbf{A}^H \boldsymbol{\alpha} + \boldsymbol{\epsilon}$$

where

$$\mathbf{A}^H = \begin{bmatrix} s_1 & 0 & \cdots & 0 & s_N & s_{N-1} & \cdots & s_2 \\ s_2 & s_1 & & \vdots & 0 & s_N & & \vdots \\ \vdots & \vdots & \ddots & 0 & \vdots & \vdots & \ddots & s_N \\ s_N & s_{N-1} & \cdots & s_1 & 0 & 0 & \cdots & 0 \end{bmatrix}$$

$$\boldsymbol{\alpha} = [\alpha_0 \ \alpha_1 \ \cdots \ \alpha_{N-1} \ \alpha_{-N+1} \ \cdots \ \alpha_{-1}]^T \in \mathbb{C}^{2N-1}.$$

In the pulse compression stage, using an MF:

$$f(\mathbf{s}) \triangleq \frac{|\mathbf{s}^H \mathbf{y}|^2}{\sum_{k \neq 0} |\mathbf{s}^H \mathbf{J}_k \mathbf{y}|^2} = \frac{\mathbf{s}^H \mathbf{A} \mathbf{s}}{\mathbf{s}^H \mathbf{B} \mathbf{s}} \triangleq \frac{n(\mathbf{s})}{d(\mathbf{s})}$$

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The optimization problem:

$$\begin{aligned} \max_{\mathbf{s}} \quad & \frac{\mathbf{s}^H \mathbf{A} \mathbf{s}}{\mathbf{s}^H \mathbf{B} \mathbf{s}} \\ \text{s.t.} \quad & |s_k| = 1, \quad k \in \{1, \dots, N\} \end{aligned}$$

Problem Formulation (Contd.)

Unimodular quadratic program (UQP):

$$\begin{aligned} \max_{\mathbf{s}} \quad & \mathbf{s}^H \boldsymbol{\chi} \mathbf{s} \\ \text{s.t.} \quad & |s_k| = 1, \quad k \in \{1, \dots, N\}. \end{aligned}$$

[1] M. Soltanalian et al. "Joint design of the receive filter and transmit sequence for active sensing", IEEE Signal Processing Letters, vol. 20, no. 5, pp. 423–426, May 2013.

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Power method like iterations (PMLI)^[1]:

$$\begin{aligned} \min_{\mathbf{s}^{(n+1)}} \quad & \left\| \mathbf{s}^{(n+1)} - \boldsymbol{\chi} \mathbf{s}^{(n)} \right\|_2, \\ \text{s.t.} \quad & |s_k^{(n+1)}| = 1, \quad \forall k \end{aligned}$$

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Analytical solution:

$$\mathbf{s}^{(n+1)} = e^{j \arg(\boldsymbol{\chi} \mathbf{s}^{(n)})}$$

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Let,

$$\tilde{g}_{\phi_i}(\mathbf{z}) = a(\mathbf{u})$$

where $\mathbf{u} = \mathbf{W}_i \mathbf{z}$, $\phi_i = \{\mathbf{W}_i\}$, and $a(\cdot)$ denotes a non-linear activation function

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Given an input \mathbf{x}_0 , **the dynamics of a fully connected DNN** with L layers:

$$\begin{aligned}\mathbf{x}_L &= \mathcal{F}(\mathbf{x}_0; \Upsilon) = \tilde{g}_{\phi_{L-1}} \circ \tilde{g}_{\phi_{L-2}} \circ \cdots \circ \tilde{g}_{\phi_0}(\mathbf{x}_0), \\ \Upsilon &= \{\phi_i\}_{i=0}^{L-1}\end{aligned}$$

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PMLI: $\mathcal{S}(\mathbf{x}) \triangleq e^{j \arg(\mathbf{x})}$

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$$\text{PMLI:} \quad \mathcal{S}(\mathbf{x}) \triangleq e^{j \arg(\mathbf{x})}$$

PMLI are perfect candidates for unfolding onto a DUN

Let,

$$g_{\phi_i}(\mathbf{z}) = \mathcal{S}(\mathbf{u})$$

where $\mathbf{u} = \boldsymbol{\chi}_i \mathbf{z}$ and $\phi_i = \{\boldsymbol{\chi}_i\}$

Deep Evolutionary Cognitive Radar (DECoR)

Let,

$$g_{\phi_i}(\mathbf{z}) = \mathcal{S}(\mathbf{u})$$

where $\mathbf{u} = \boldsymbol{\chi}_i \mathbf{z}$ and $\phi_i = \{\boldsymbol{\chi}_i\}$

The dynamics of DECoR for L layers:

$$\mathbf{s}_L = \mathcal{G}(\mathbf{s}_0; \boldsymbol{\Omega}) = g_{\phi_{L-1}} \circ g_{\phi_{L-2}} \circ \cdots \circ g_{\phi_0}(\mathbf{s}_0)$$

Deep Evolutionary Cognitive Radar (DECoR)

Let,

$$g_{\phi_i}(z) = \mathcal{S}(u)$$

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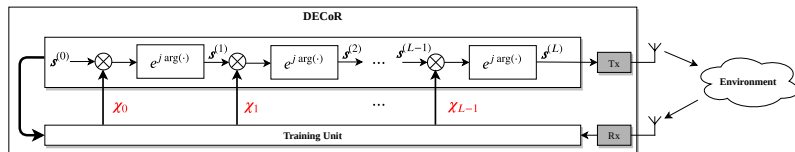


Figure: The proposed DECoR architecture for adaptive radar waveform design

Step 0: (Initialization): Choose unimodular transmit sequence $\mathbf{s}_0 \in \mathbb{C}^N$, set $t = 0$, $\sigma = c$, $\delta \in (0, 1]$, $\Omega^{(0)} = \{\mathbf{x}_i^{(0)}\}_{i=0}^{L-1}$ such that $\mathbf{x}_i^{(0)} \succ 0$, for $i \in \{0, \dots, L-1\}$.

Step 1: (Random walk- generation): For $l \in \{0, \dots, L-1\}$ and $i \in \{0, \dots, B-1\}$ generate $\mathbf{D}_l^i \in \mathbb{C}^{N \times N}$ as the set of Hermitian positive-definite search direction matrices.

Step 2: (Random walk- perturbation): For $i \in \{0, \dots, B-1\}$, form the parameter space $\Omega_i^{(t)}$ as $\Omega_i^{(t)} = \{\mathbf{x}_0^{(t)} + \mathbf{D}_0^i, \dots, \mathbf{x}_{L-1}^{(t)} + \mathbf{D}_{L-1}^i\}$. Compute $\mathbf{s}_{L,i}^{(t)} = \mathcal{G}(\mathbf{s}_0; \Omega_i^{(t)})$ for $i \in \{0, \dots, B\}$ and $\mathbf{S}^{(t)} = \{\mathbf{s}_{L,0}^{(t)}, \dots, \mathbf{s}_{L,B-1}^{(t)}\}$.

Step 3: (Collecting information): Transmit $\mathbf{S}^{(t)}$ and obtain $\mathbf{Y} = \{\mathbf{y}_0^{(t)}, \dots, \mathbf{y}_{B-1}^{(t)}\}$. Compute $f(\mathbf{s})$ for each transmit/receive pair $(\mathbf{s}_{L,i}^{(t)}, \mathbf{y}_i^{(t)})$ and construct $\mathcal{F} = \{f(\mathbf{s}_{L,i}^{(t)})\}_{i=0}^{B-1}$.

Step 4: (Optimizing the DECoR architecture): Choose i_*

$$i_* = \arg \max_{i \in [B]} f(\mathbf{s}_{L,i}^{(t)}).$$

Update the network parameters if $f(\mathbf{s}_{L,i_*}^{(t)}) \geq f(\mathbf{s}_{L,i_*}^{(t-1)})$ and set $\sigma \leftarrow c$. Otherwise, only update the search radius as $\sigma \leftarrow \delta\sigma$. Continue the online learning by going to Step 1.

Numerical Examples

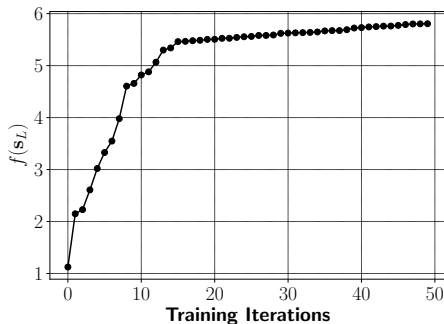


Figure: The objective value $f(s_L)$ of the DECoR vs. training iterations for a code length of $N = 10$

Numerical Examples (contd.)

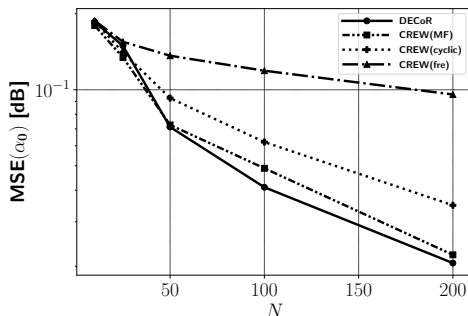


Figure: MSE values obtained by the different design algorithms for code lengths $N \in \{10, 25, 50, 100, 200\}$.

- In order to design cognitive radar waveforms, we **bridge the gap between model-based and data driven techniques** and propose a methodology to unfold the PMLI to solve a UQP
- Although the **DECoR framework does not have access to the statistics** of the environmental parameters, it is able to **learn them by exploiting the observed data** from interaction with the environment.

Thank you
and
Questions?